

Applications of concentration of measure in signal processing

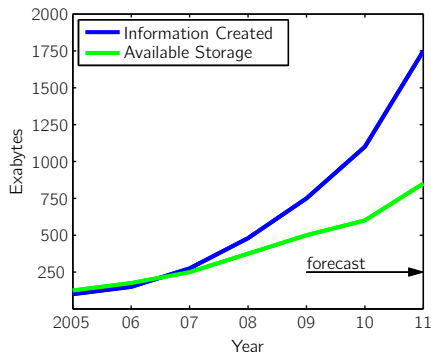
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April 30, 2010

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 - Sparse Signal Models
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Signal Processing in the Age of the Data Flood

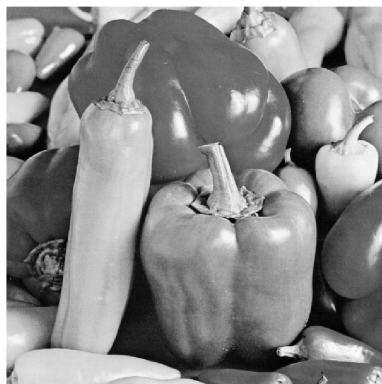
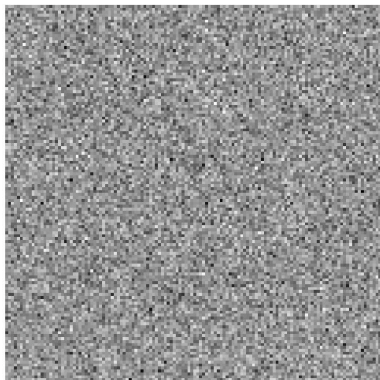


Reference: Economist magazine,
Feb 25, 2010.

- Exabyte = 2^{60} bits.
- We have passed the point where all data created can be stored
- LHC generates 40 Tb every second.
- Other bottlenecks
 - acquisition
 - transmission
 - analysis

Not all length- N signals are created equal

- What is the class of “typical images”?




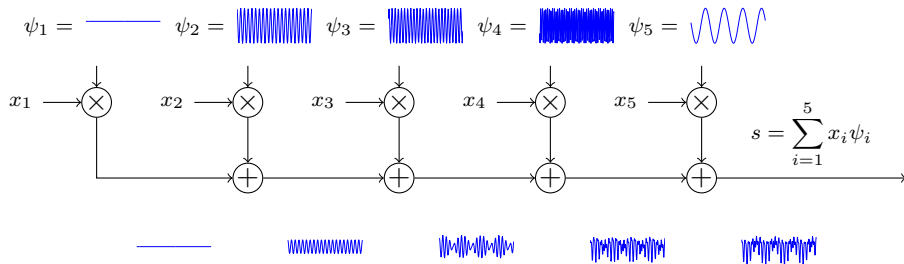
- “Typical” signals contain degrees of freedom S less than N

Dimensionality Reduction

- Can we reduce the burden from N to S early and often in the data processing pipeline?

Signal Representation: Signal Basis

- A signal basis can be used to define the class of signals of interest
- Example: represent a signal $z =$  as sum of scaled sinusoids



Lossy Compression: JPEG

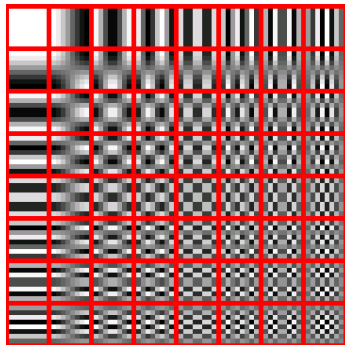


8×8 block



$$z = \sum_{n=1}^{64} x_n \psi_n$$

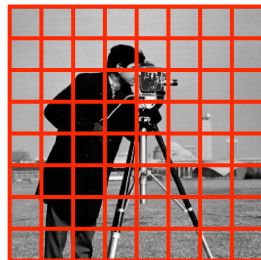
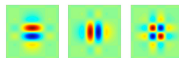
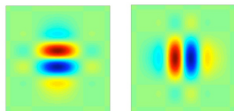
Discrete Cosine Signal Basis ψ_k



credit: J. Romberg

- Approximation with quantized coeffs: $\hat{z} = \sum_{n=1}^{64} \hat{x}_n \psi_n;$

Multi-Scale Basis: Wavelets

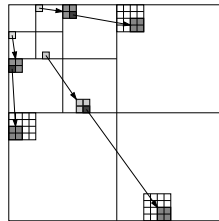
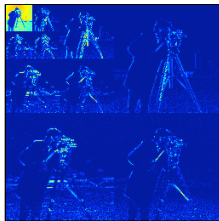


credit: J. Romberg

Wavelet coefficient representation



credit: J. Romberg



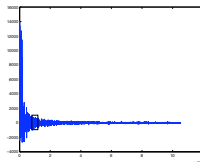
- A few large coefficients, but many small coefficients.

How many coefficients are important?

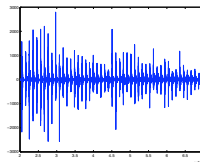
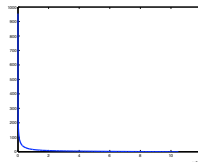


1 megapixel image

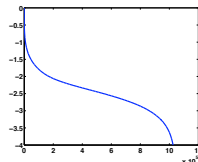
wavelet coeffs



(sorted)



zoom in

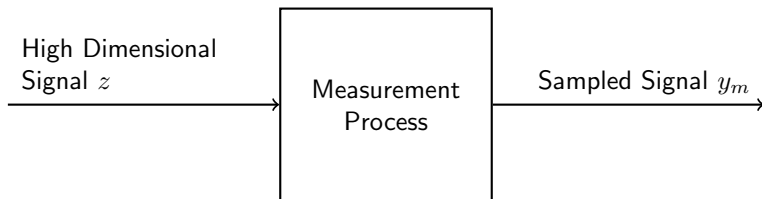


(log₁₀ sorted)

Conclusion

- Many classes of signals have a *sparse* representation in an appropriate basis

Now let's bring in the measurement



Measurement Models

- Many measurement modalities are *linear*
- Inner product representation:

$$y_m = \langle z, \phi_m \rangle = \text{sum of point-wise product}$$

- Tomography

$$y_m = \left\langle \text{Image of person with camera on tripod}, \text{Edge detector} \right\rangle$$

credit: J. Romberg

Band-limited Signal Recovery:

Signal and Measurement Model

- The signal/measurement model for the (1-D) Nyquist theorem uses
 - Signal model basis:

$$\psi_n = e^{j\omega_0 n t}$$

- Measurement: Sampling, M samples per period $T_s = T_0/M$.

$$\phi_m = \delta(t - T_s m)$$

Sampling frequency is $f_s = \frac{1}{T_s} = \frac{M}{T_0} = M\omega_0$.

- *a priori* information: Band-limited signal, i.e. coefficients zero for $|n| \geq N_b$. Bandwidth: $\omega_b = N_b\omega_0$

Band-limited Signal Recovery:

Set of Linear Equations

- Using this model,

$$y_m = \left\langle \sum_{n=-N_b}^{N_b} x_n \psi_n, \phi_m \right\rangle$$

$$y_m = \begin{bmatrix} a'_m \end{bmatrix} \begin{bmatrix} x_{-N_b} \\ x_{1-N_b} \\ \vdots \\ x_{N_b-1} \\ x_{N_b} \end{bmatrix}$$

where $a_k = [\langle \psi_{-N_b}, \phi_m \rangle \cdots \langle \psi_{N_b}, \phi_m \rangle]$

Band-limited Signal Recovery:

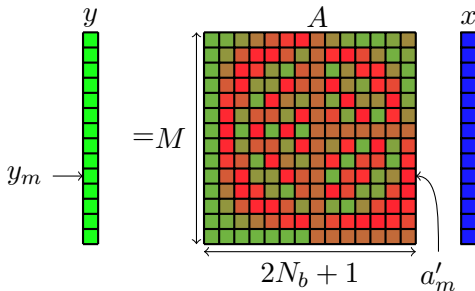
Row Independence

$$\begin{aligned} a_k &= [\langle \psi_{-N_b}, \phi_m \rangle \quad \cdots \quad \langle \psi_{N_b}, \phi_m \rangle] \\ &= [e^{j\omega_0 T_s m(-N_b)} \quad e^{j\omega_0 T_s m(-N_b+1)} \quad \cdots \quad e^{j\omega_0 T_s m(N_b)}] \end{aligned}$$

- a_k looks like $e^{j\hat{\omega}n}$ with $\hat{\omega} = \omega_0 T_s m = \frac{2\pi}{M}$
- Orthogonality property of complex exponentials: a_i and a_j are orthogonal (and thus independent) for $0 < i \neq j \leq M$.

Band-limited Signal Recovery:

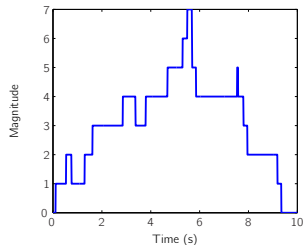
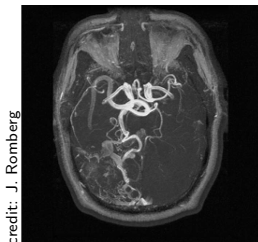
Nyquist Recovery



- Since rows are independent, need $M \geq 2N_b + 1$ to recover x .
- Implies $f_s \geq (2N_b + 1)\omega_0$: sampling frequency needs to be greater than two times bandwidth.

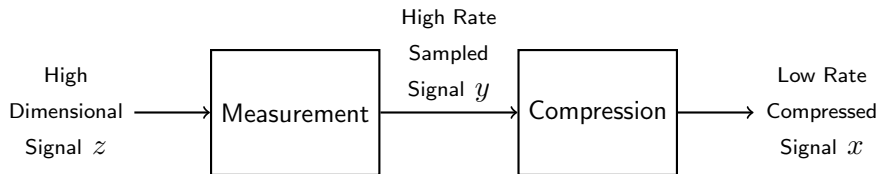
So what is the problem?

- Signals often have high bandwidth, but lower complexity content



- What if we change the signal model: not bandlimited, but sparse in some basis.

Current Solution: Measure Then Compress



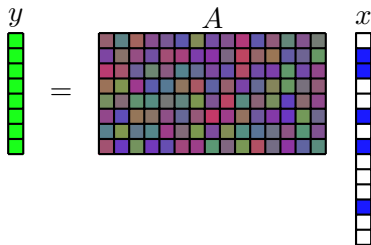
- Measurement costs \$. Compression costs \$.
- Can we combine the measurement and compression steps? (Compressive Sensing)

Sparse Signal Recovery:

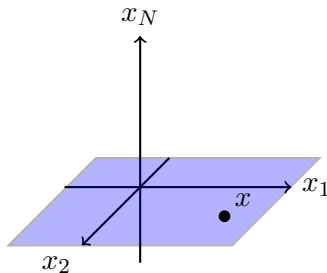
Compressive Measurement Model

- Model: signal $x \in \mathbb{R}^N$, with S -sparse support, measurement, $y \in \mathbb{R}^M$.
 - Ψ - signal basis (columns are ψ_n)
 - Φ - measurement matrix (rows are ϕ_m)

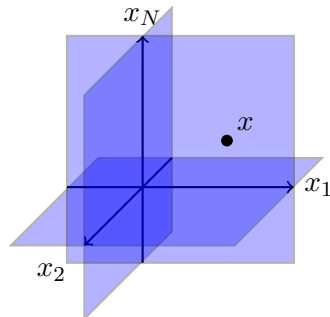
$$y = \underbrace{\Phi \Psi}_A x$$



Geometry of Signal Models



Linear Subspace, $\dim N_b$
Bandlimited Signals



Union of $\dim S$ Subspaces
Sparse Signals

Sparse Signal Recovery:

Recovery via regularization

- Given y , can we recover x ?
- A is short and fat: non-trivial null space means many solutions to $y = Ax$.
- Idea: regularized recovery

$$\hat{x} = \arg \min_x \|x\|_* \quad \text{s.t. } y = Ax$$

Sparse Signal Recovery:

ℓ_2 recovery

- ℓ_2 -recovery (Euclidian distance) doesn't work

$$\hat{x} = \arg \min_x \|x\|_2 \quad \text{s.t. } y = Ax$$

- Minimum is almost never sparse



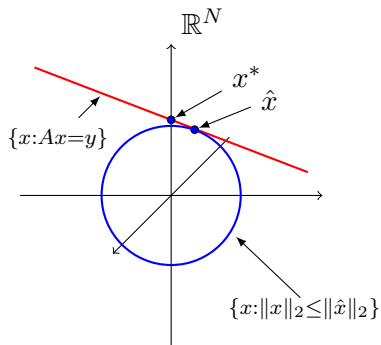
x



$$\hat{x} = (A'A)^{-1}A'y$$

Sparse Signal Recovery:

ℓ_2 recovery geometry



Incorrect Recovery

Sparse Signal Recovery:

Sparcity preserving norms

- ℓ_0 -recovery: $\|x\|_0 = \#$ of non-zero elements of x .

$$\hat{x} = \arg \min_x \|x\|_0 \quad \text{s.t. } y = Ax$$

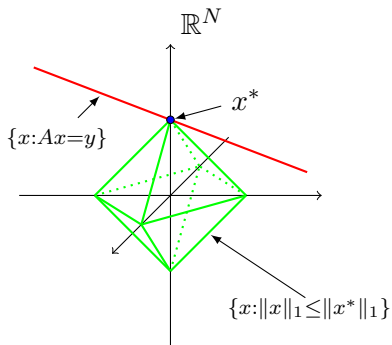
- Works generically if $M = S + 1$. However, computationally demanding.
- ℓ_1 -recovery: $\|x\| = \sum |x_i|$. Convex! Recovery via LP:

$$\hat{x} = \arg \min_x \|x\|_1 \quad \text{s.t. } y = Ax$$

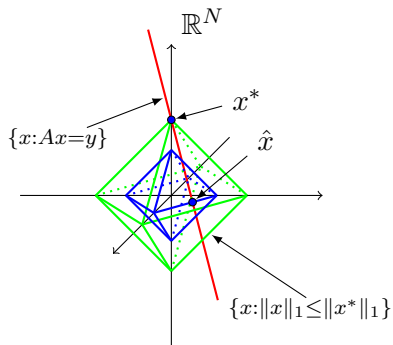
- Also related to basis pursuit, lasso.
- Works generically if $M \approx S \log N!!!$

Sparse Signal Recovery:

ℓ_1 recovery geometry



Correct Recovery



Incorrect Recovery

Other Recovery Methods

- Greedy methods - Orthogonal Matching Pursuit (Tropp, 2004)
- Iterative convex - Reweighted ℓ_1 - (Candès, Wakin and Boyd, 2008)
- Non-convex - smoothed ℓ_0 - (Chartrand, 2007; Mohimani et al., 2007)

Recovery Example

256x256
original



6500 wavelets



26000 random
projections

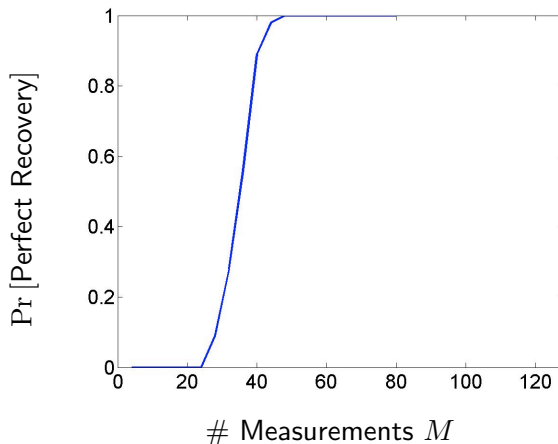


credit: J. Romberg

- Wavelets: 6500 largest coefficients
- 26000 random projections: recovery using wavelet basis
- Good approximation with 4x sampling rate over perfect knowledge

Recovery Example

Signal length $N = 128$, $S = 10$



Sparse Signal Detection

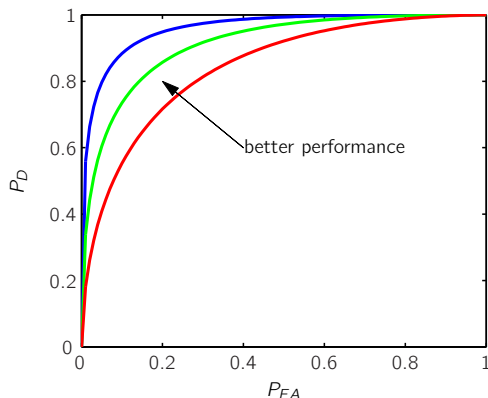
- The classic signal detection problem:
 - known signal z may or may not have been sent
 - measurement y corrupted by noise v
- Define events \mathcal{E}_0 and \mathcal{E}_1 as:

$$\mathcal{E}_0 \triangleq y = v$$

$$\mathcal{E}_1 \triangleq y = z + v$$

- Detection algorithm: decide if event \mathcal{E}_0 or \mathcal{E}_1 occurred.
- Performance metrics are
 - false-alarm probability - $P_{FA} = \Pr[(\mathcal{E}_1 \text{ chosen when } \mathcal{E}_0)]$
 - detection probability - $P_D = \Pr[(\mathcal{E}_1 \text{ chosen when } \mathcal{E}_1)]$

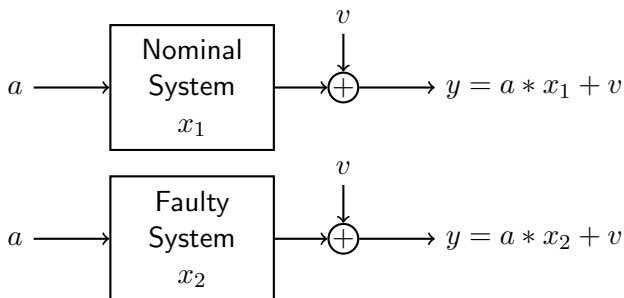
Receiver Operation Characteristic: ROC curve



- How many measurements are necessary to obtain the desired performance?

Fault Isolation

- Application: System with known fault condition. All signals are discrete time sequences.



- Subtract expected output: detection problem with $z = a * (x_1 - x_2)$
- Convolution: $z = A(x_1 - x_2)$, A Toeplitz Matrix.

Compressive Signal Processing

- Experiment model

$$y = \Phi\Psi x + v$$

- Ψ - signal basis (columns are ψ_n)
- Φ - measurement matrix (rows are ϕ_m)
- $y \in \mathbb{R}^M$, measurement, $x \in \mathbb{R}^N$, S -sparse signal, $v \in \mathbb{R}^M$ measurement noise.

- Basic problems

- Compressive recovery of unknown S -sparse signal using M measurements, with $S < M \ll N$.
- Detection of a known S -sparse signal using M measurements, with $S < M \ll N$.

Compressive Signal Processing:

Questions

- What are the conditions that guarantee that all x of a given sparsity can be recovered?
- What are the conditions that guarantee a particular level of performance in detection?
- How can we generate measurement matrices that meet these conditions?

The Restricted Isometry Property (RIP)

- Introduced by Candès and Tao

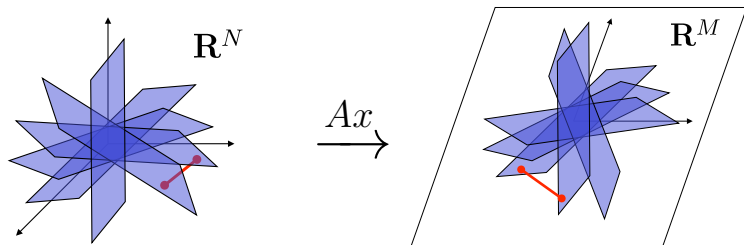
Definition

X satisfies the RIP of order S if there exists a $\delta_S \in (0, 1)$ such that

$$(1 - \delta_S) \|a\|_2^2 \leq \|Xa\|_2^2 \leq (1 + \delta_S) \|a\|_2^2$$

holds for all S -sparse signals a .

RIP as embedding



- Difference of two S -sparse signals is $2S$ sparse.

$$(1 - \delta_{2S}) \|u - v\|_2^2 \leq \|A(u - v)\|_2^2 \leq (1 + \delta_{2S}) \|u - v\|_2^2$$

Recovery Result:

Candès (2008)

- Recovery algorithm (basis pursuit de-noising)

$$\hat{x} = \arg \min_x \|x\|_1 \quad \text{s.t.} \quad \|y - Ax\|_2 \leq \epsilon$$

Theorem

Suppose y is generated by $y = Ax^ + v$. If A satisfies RIP with $\delta_{2S} < \sqrt{2} - 1$ and $\|v\|_2 < \epsilon$, then*

$$\|\hat{x} - x^*\|_2 \leq C_0 \frac{\|x^* - x_s\|}{\sqrt{s}} + C_1 \epsilon$$

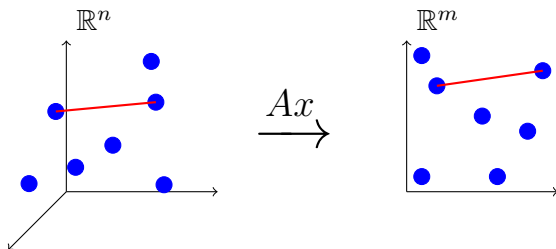
where x_s is the S -sparse approximation of x^ .*

- Implies perfect recovery if x^* is S -sparse and no noise.

Checking RIP

- Given A , does it satisfy RIP?
 - Check eigenvalues of each $M \times S$ submatrix - combinatorial.
- Generate A *randomly* - satisfies RIP with high probability when $M = O(S \log N)$!
 - iid Gaussian entries
 - iid Bernoulli entries ($+/- 1$)
 - random Fourier ensemble
 - (Candes, Tao; Donoho; Traub, Wozniakowski; Litvak et al)
- Proofs bound eigenvalues of random matrices, but generally difficult to generalize to $\Psi \neq I$.

Recall Johnson-Lindenstrauss Embedding



J-L Embedding

Given $\epsilon > 0$ and set \mathbb{P} of P points in \mathbb{R}^N , find A such that for all $u, v \in \mathbb{P}$,

$$(1 - \epsilon)\|u - v\|^2 \leq \|A(u - v)\|^2 \leq (1 + \epsilon)\|u - v\|^2$$

Random J-L Embeddings

- Using our results from the last talk, we have the following:

Theorem (Dasgupta and Gupta; Frankl; Achioptas; Indyk and Motwani)

Given set \mathbb{P} of P points in \mathbb{R}^N , choose $\epsilon > 0$ and $\beta > 0$. Let A be an $M \times N$ matrix with independent elements $[A]_{ij} \sim \mathcal{N}(0, \frac{1}{M})$ where

$$M \geq \left(\frac{7 + 6\beta}{\min(.5, \epsilon)^2} \right) \ln(P).$$

Then with probability greater than $1 - P^{-\beta}$, the following holds: For all $u, v \in \mathbb{P}$,

$$(1 - \epsilon)\|u - v\|^2 \leq \|A(u - v)\|^2 \leq (1 + \epsilon)\|u - v\|^2$$

Other Favorable Random Mappings:

Sub-Gaussian Distributions

- In the proof, we used

$$[A]_{ij} \sim \mathcal{N}\left(0, \frac{1}{M}\right)$$

- Key step was Chernoff bound using moment generating function

Definition

A random variable X is *Sub-Gaussian* if there exists an $a \geq 0$ such that

$$\mathbb{E} \left[e^{sX} \right] \leq e^{\frac{a^2 s^2}{2}}$$

and τ , the smallest such a , is called the *Gaussian standard* of X .

Other Favorable Random Mappings:

Properties of Sub-Gaussians

Key Properties

- If X_i are iid sub-Gaussian, $Y = \sum X_i$ is sub-Gaussian with standard $\tau_y \leq \sum \tau_{x_i}$
- If X is sub-Gaussian with standard τ , $\mathbb{E} \left[e^{sX^2} \right] \leq \frac{1}{1-2s\tau^2}$

Other Favorable Random Mappings:

Sub-Gaussian Examples

- We can use any zero mean sub-Gaussian iid sequence with variance $1/M$.
- Rademacher Sequence

$$[A]_{ij} = \begin{cases} +\frac{1}{\sqrt{M}} & \text{with probability } \frac{1}{2} \\ -\frac{1}{\sqrt{M}} & \text{with probability } \frac{1}{2} \end{cases}$$

- “Database-friendly” (Achlioptas)

$$[A]_{ij} = \begin{cases} +\sqrt{\frac{3}{M}} & \text{with probability } \frac{1}{6} \\ 0 & \text{with probability } \frac{1}{3} \\ -\sqrt{\frac{3}{M}} & \text{with probability } \frac{1}{6} \end{cases}$$

From JL to RIP

- Baraniuk et al. (2008)
- Consider measurement with $\Psi = I$, Φ random elements from a favorable distribution

$$y = \underbrace{\Phi\Psi}_A x$$

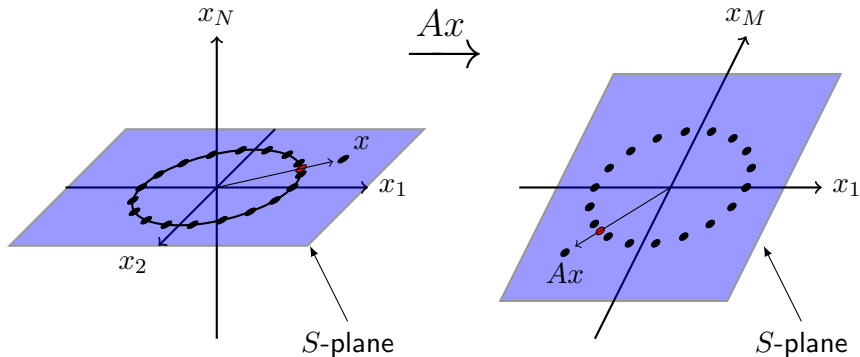
- Favorable distribution implies that for given $x \in \mathbb{R}^N$,

$$\Pr \left[\left| \|Ax\|_2^2 - \|x\|_2^2 \right| \geq \epsilon \|x\|_2^2 \right] \leq 2e^{-Mc_0(\epsilon)}$$

- pick $\epsilon = \delta_{2S}/2$

From JL to RIP

- Examine mapping on one of $\binom{N}{S}$ S -planes in sparse model
 - Construct (careful) covering of unit sphere using $(12/\delta_{2S})^S$ points
 - JL: isometry for each point with high probability
 - Union bound for all points
 - Extend isometry to all x in unit ball (and thus all x in S -plane)



A look at the probabilities:

Union Bounds

- Probability of error $> \frac{\delta_{2S}}{2}$ when mapping 1 point

$$\leq 2e^{-Mc_0(\delta_{2S}/2)}$$

- Probability of error when $(12/\delta_{2S})^S$ points mapped

$$\leq 2(12/\delta_{2S})^S e^{-Mc_0(\delta_{2S}/2)}$$

- “Careful” covering implies that for all x in unit ball, $\exists q$ in covering s.t. $\|x - q\| < \delta_{2S}/4$.
- Probability of error $> \delta_{2S}$ when unit ball mapped

$$\leq 2(12/\delta_{2S})^S e^{-Mc_0(\delta_{2S}/2)}$$

A look at the probabilities, continued

- Probability of error $> \delta_{2S}$ when $\binom{N}{S}$ planes mapped:

$$\leq 2 \binom{N}{S} (12/\delta_{2S})^k e^{-Mc_0(\delta_{2S}/2)} \leq 2e^{-c_0(\delta_{2S})M + S[\ln(eN/S) + \ln(12/\delta_{2S})]}$$

Result

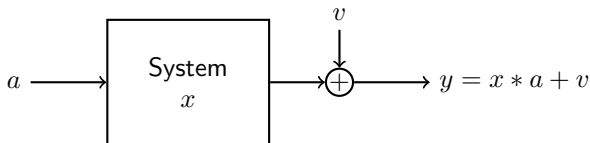
If $M > O(S \log(N/S))$, with probability greater than $1 - 2e^{-c_2M}$, A random matrix with favorable distribution satisfies RIP.

- Bonus: Universality for orthonormal basis Ψ : only changes orientation of planes in model.

Structured Measurements:

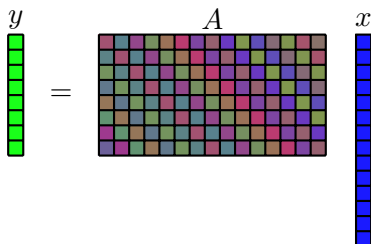
A Detection Problem with Convolution

- We are not always free to choose the elements of Φ independently
 - Distributed measurements
 - Dynamic Systems



Convolution implies Toeplitz measurement matrix

$$y = a * x$$



- Cannot choose the elements of A independently

Concentration of Measure for Toeplitz matrices

- Suppose a is chosen iid Gaussian $x_i \sim \mathcal{N}(0, \frac{1}{M})$.
- For fixed $x, y \sim \mathcal{N}(0, \frac{1}{M}P)$ where

$$[P]_{ij} = \sum_{i=1}^{n-|i-j|} x_i x_{i+|i-j|}$$

- Let $\rho(x) = \frac{\lambda_{\max}(P)}{\|x\|_2^2}$ and $\mu(x) = \frac{\frac{1}{d} \sum \lambda_i^2(P)}{\|x\|_2^2}$.

Result

For any $\epsilon \in (0, 0.5)$

$$\Pr [\|Ax\|_2^2 \geq \|x\|_2^2 (1 + \epsilon)] \leq e^{-\epsilon^2 M / 6\rho(a)}$$

$$\Pr [\|Ax\|_2^2 \leq \|x\|_2^2 (1 - \epsilon)] \leq e^{-\epsilon^2 M / 4\mu(a)}$$

Implications

- Result for A Toeplitz:

$$\Pr [\|Ax\|_2^2 \geq \|x\|_2^2 (1 + \epsilon)] \leq e^{-\epsilon^2 M / 6 \rho(a)}$$

$$\Pr [\|Ax\|_2^2 \leq \|x\|_2^2 (1 - \epsilon)] \leq e^{-\epsilon^2 M / 4 \mu(a)}$$

- Recall result from previous lecture for A unstructured:

$$\Pr [\|Ax\|_2^2 \geq \|x\|_2^2 (1 + \epsilon)] \leq e^{-\epsilon^2 M / 6}$$

$$\Pr [\|Ax\|_2^2 \leq \|x\|_2^2 (1 - \epsilon)] \leq e^{-\epsilon^2 M / 4}$$

- Concentration bound worsens over i.i.d. entries by factors ρ and μ .
- Bound: $\mu(a) \leq \rho(a) \leq \|a\|_0$. However, most a are much less than this bound.

Fault Detection Problem

- System impulse response can be x_1 or x_2 .
- record $\tilde{y} = y - Ax_1$, let $\delta x = x_2 - x_1$
- Define events \mathcal{E}_0 and \mathcal{E}_1 as:

$$\mathcal{E}_0 \triangleq \tilde{y} = v$$

$$\mathcal{E}_1 \triangleq \tilde{y} = A\delta x + v$$

- Detection algorithm: decide if event \mathcal{E}_0 or \mathcal{E}_1 occurred.
- Performance metrics are
 - false-alarm probability - $P_{FA} = \Pr[(\mathcal{E}_1 \text{ chosen when } \mathcal{E}_0)]$
 - detection probability - $P_D = \Pr[(\mathcal{E}_1 \text{ chosen when } \mathcal{E}_1)]$

Neyman-Pearson Test

- The Neyman-Pearson detector maximizes P_D for a given limit on failure probability, $P_{FA} \leq \alpha$ under Gaussian noise assumption.

$$\tilde{y}' Ax \underset{\mathcal{E}_0}{\overset{\mathcal{E}_1}{\geq}} \gamma$$

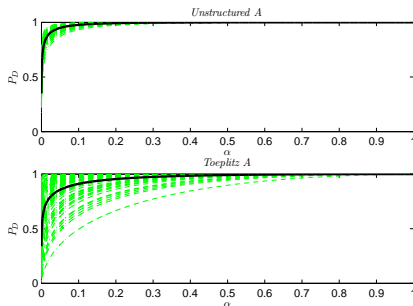
- Performance:

$$P_D = Q \left(Q^{-1}(P_{FA}) - \frac{\|Ax\|_2}{\sigma} \right)$$

- Since performance depends on $\|Ax\|_2$, worse performance for signals with large $\rho(a)$, $\mu(a)$.

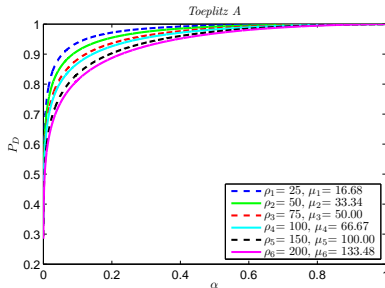
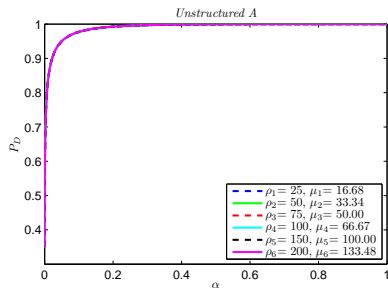
Detection Performance

- $y = Ax + v$
- A is 125×250
- x is block sparse,
 $\mu(a) = 33, \rho(a) = 50$.
- Two cases:
 - A - Unstructured
 - A - Toeplitz
 - 1000 realizations of A



Detection Performance

- Average detection performance for six different x .



Conclusion

- Compressive Sensing - going beyond Nyquist sampling
- Sparse signal model with linear measurement model
- Recovery possible using convex optimization
- Work continues on
 - Recovery methods
 - Structured measurements
 - New applications - development of sparse signal models
 - ...